

# Final Exam

EEL 3657 (Summer 2003)

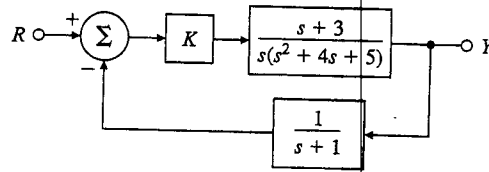
Name: Solutions

Student ID#: \_\_\_\_\_

## Instructions

1. There are totally 4 problems in this exam.
2. Show all work for partial credit.
3. Assemble your work for each problem in logical order.
4. Justify your conclusion. I cannot read minds.

**Problem 1.** (25 points) Consider the system shown below,



- 1). Find the closed-loop characteristic equation.
- 2). Use Routh's stability criterion, determine all values of K for which the system is stable.

Solution: 1). 
$$G_{cl}(s) = \frac{K \cdot \frac{s+3}{s(s^2+4s+5)}}{1 + K \cdot \frac{s+3}{s(s^2+4s+5)} \cdot \frac{1}{s+1}}$$

$$= \frac{K(s+1)(s+3)}{s(s^2+4s+5)(s+1) + K(s+3)}$$

$$= \frac{Ks^2 + 4Ks + 3K}{s^4 + 5s^3 + 9s^2 + (5+K)s + 3K}$$

CE: 
$$s^4 + 5s^3 + 9s^2 + (5+K)s + 3K = 0$$

2). Routh table:

$s^4$	1	9	$3k$
$s^3$	5	$5+k$	
$s^2$	$b_1$	$b_2$	
$s^1$	$c_1$	$c_2$	
$s^0$	$d_1$		

$$b_1 = \frac{40-k}{5}$$

$$b_2 = 3k$$

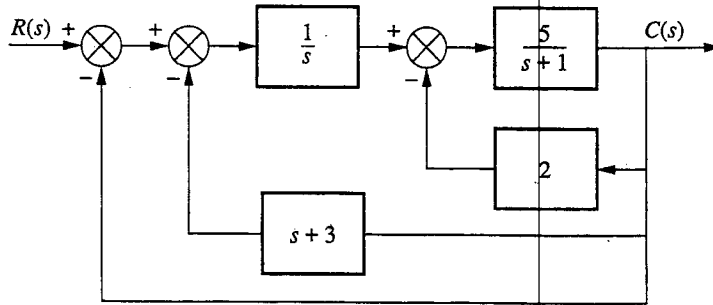
$$c_1 = (5+k) - \frac{75k}{40-k}$$

$$c_2 = 0$$

$$d_1 = 3k$$

$$\left. \begin{array}{l} b_1 > 0 \\ c_1 > 0 \\ d_1 > 0 \end{array} \right\} \Rightarrow \left\{ \begin{array}{l} 40-k > 0 \Rightarrow k < 40 \\ (5+k) - \frac{75k}{40-k} > 0 \Rightarrow -4.49 < k < 4.49 \\ 3k > 0 \Rightarrow k > 0 \end{array} \right\} \Rightarrow 0 < k < 4.49$$

**Problem 2.** (25 points) For the system shown below, what steady-state error can be expected for an input of  $15u(t)$ ? ( $u(t)$  is a unit step function.)



Solution:

Reduce the system to an equivalent unity feedback system.

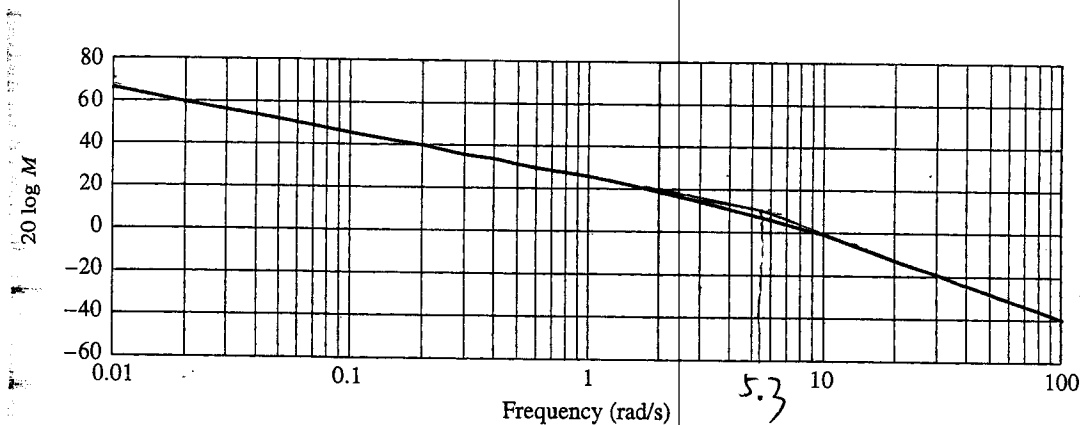
$$G(s) = \frac{\frac{5}{s(s+11)}}{1 + \frac{5(s+3)}{s(s+11)}} = \frac{5}{s(s+11) + 5(s+3)} = \frac{5}{s^2 + 16s + 15}$$

Hence,  $e(\infty) = \lim_{s \rightarrow 0} G(s) = K_p = \frac{1}{3}$ .

Finally,

$$e_{ss} = \frac{15}{1 + K_p} = \frac{15}{1 + \frac{1}{3}} = \frac{45}{4} = 11.25$$

**Problem 3.** (20 points) For the Bode magnitude plot (with straight-line segments marked) shown below, determine the transfer function.



Solution: The first <sup>(left)</sup> segment has a slope  $-20$  dB/dec,  
which indicates a  $\frac{1}{s}$  term;

The second <sup>(right)</sup> segment has a slope  $-40$  dB/dec,  
which indicates a  $\frac{1}{1 + \frac{s}{5.3}}$  term.

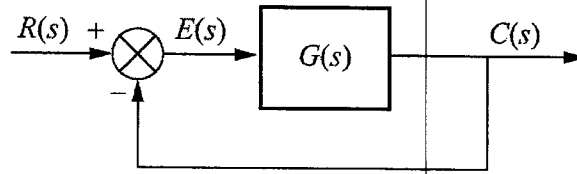
Determine  $K$  from low-frequency Bode magnitude plot,

$$67 = 20 \log k - 20 \log 0.01$$

$$\Rightarrow k = 22.39$$

$$\text{So } G(s) = \frac{k}{s(1 + \frac{s}{5.3})} = \frac{5.3k}{s(s+5.3)} = \frac{118}{s(s+5.3)}$$

**Problem 4.** (30 points) For a unity feedback system shown below,



where  $G(s) = \frac{K}{s(s+2)(s+10)}$ ,

- 1). Sketch the Nyquist plot for  $K=1$ , and use Nyquist criterion to claim stability for the closed-loop system at  $K=1$ .
- 2). For what range of  $K$  is the closed-loop system stable?
- 3). What is the gain margin?

Solution: 1).

1. Substitute  $s = j\omega$  in  $G(s)$   
Setting  $s = j\omega$  we get

$$G(j\omega) = \frac{1}{j\omega(j\omega + 2)(j\omega + 10)}$$

2. Substituting  $\omega = 0$  in the last equation, we get the zero-frequency property of  $L(j\omega)$ ,

$$G(j0) = \infty \angle -90^\circ$$

3. Substituting  $\omega = \infty$ , the property of the Nyquist plot at infinite frequency is established.

$$G(j\infty) = 0 \angle -270^\circ$$

4. To find the intersect(s) of the Nyquist plot with the real axis, if any, we rationalize  $G(j\omega)$  by multiplying the numerator and the denominator of the equation by the complex conjugate of the denominator. Thus, Eq. (9-59) becomes

$$\begin{aligned} G(j\omega) &= \frac{[-12\omega^2 - j\omega(20 - \omega^2)]}{[-12\omega^2 + j\omega(20 - \omega^2)][-12\omega^2 - j\omega(20 - \omega^2)]} \\ &= \frac{[-12\omega^2 - j\omega(20 - \omega^2)]}{\omega[144\omega^2 + (20 - \omega^2)]} \end{aligned}$$

5. To find the possible intersects on the real axis, we set the imaginary part of  $G(j\omega)$  to zero. The result is

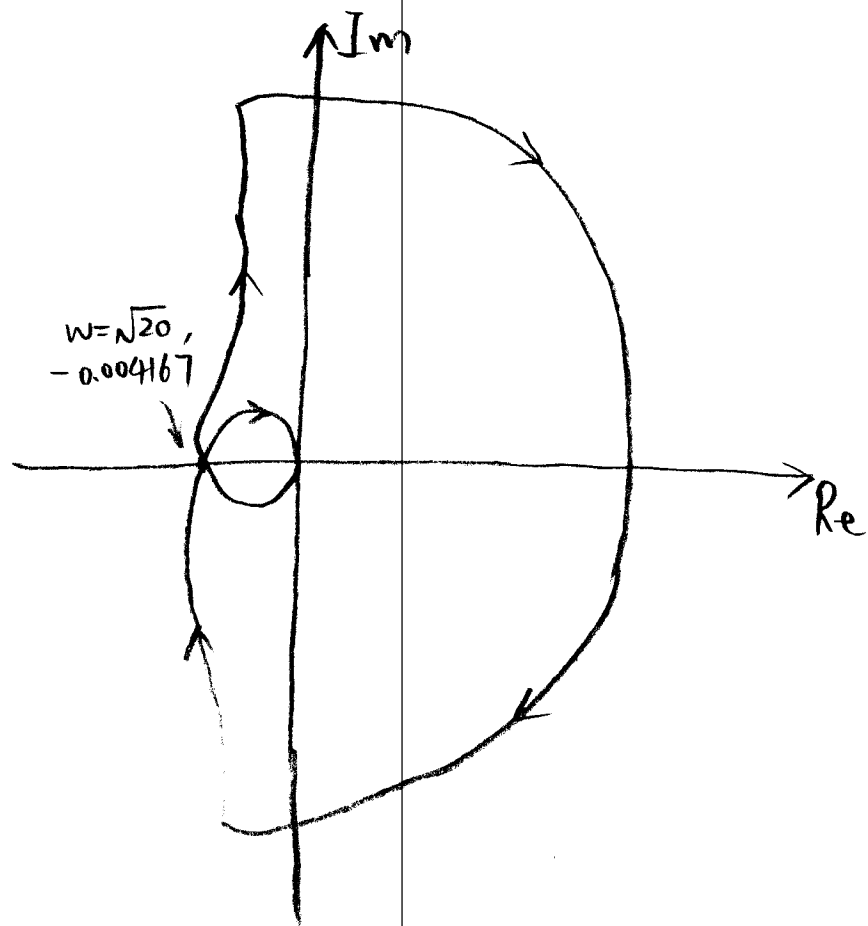
$$\text{Im}[G(j\omega)] = \frac{-(20 - \omega^2)}{\omega[144\omega^2 + (20 - \omega^2)]} = 0$$

The solutions of the last equation are:  $\omega = \infty$ , which is known to be a solution at  $G(j\omega) = 0$ , and

$$\omega = \pm \sqrt{20} \text{ rad/sec}$$

Since  $\omega$  is positive, the correct answer is  $\omega = \sqrt{20}$  rad/sec. Substituting this frequency, we have the intersect on the real axis of the  $G(j\omega)$ -plane at

$$G(j\sqrt{20}) = -\frac{12}{2880} = -0.004167$$



2). 
$$k = \frac{1}{0.004167} = 240$$

3). gain margin = 240